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Real-time Damper Force Estimation for Automotive Suspension

A Generalized H₂/LPV Approach

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Abstract

The real-time knowledge of the damper force is of paramount importance in controlling and diagnosing automotive suspension systems. This study presents a generalized H_2 /LPV observer for damper force estimation of a semi-active electro-rheological (ER) suspension system. First, an extended quarter-car model augmented with the nonlinear and dynamical model of the semi-active suspension system is written into the quasi-LPV formulation. Then, the damper force estimation method is developed through a generalized H_2 /LPV observer whose objective is to handle the impact of unknown road disturbances and sensor noise on the estimation errors of the state variables thanks to the H_2 norm. The measured sprung and unsprung mass accelerations of the quarter-car system are used as inputs for the observer. The proposed approach is simulated with validated model of the 1/5-scaled real vehicle testbed of GIPSA-lab. Simulation results show the performance of the estimation method against unknown disturbances, emphasizing the effectiveness of the damper force estimation in real time.

Keywords

semi-active suspension, damper force estimation, generalized H₂/LPV observer, quasi-LPV

1 Introduction

Semi-active suspension systems remain an interesting research topic for both academia and industry thanks to their advantages, including lower weight and energy consumption compared to active and passive ones (see Savaresi et al., 2010). In their application, the real-time knowledge of damper (or damping) force plays a vital role in controlling (see Do et al. (2010); Nguyen et al. (2015); Poussot-Vassal et al. (2008; 2012) and Priyandoko et al. (2009)) and monitoring these systems (see Morato et al. (2019)). However, the direct measurement of this force using sensors faces some issues such as difficulty and expenses in installing and maintaining these sensors, leading to the increasing demand in adopting observers to estimate the damper force. The main requirements of a damper force estimation approach include considering the dynamic characteristics of the semi-active damper and handling the nonlinearity and the unknown inputs in the dynamic model, as well as the use of the low-cost sensors (Pham, 2020). According to the considered dynamic behavior of the semi-active damper, the estimation methodologies are classified into two categories. In the first one, the parallel Kalman filters (Koch et al., 2010) and the data analysis methods (Savaresi et al., 2019) were developed to estimate the damper force, ignoring the dynamic characteristics of the semi-active damper. The second category is developed based on the dynamic damper models (Tudon-Martinez et al., 2018; Pham et al., 2019a; 2019b; 2019c;

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Vela et al., 2018). From the viewpoint of the required sensors, Tudon-Martinez et al. (2018) and Vela et al. (2018) presented a H_{∞} observer and a linear parameter-varying (LPV) - H_x filter, respectively, for damper force estimation using the deflection and the deflection velocity as inputs. However, it is worth noting that deflection sensors are also costly and difficult to install in commercial cars. Therefore, based on accelerometers, some estimation approaches are developed by Koch et al. (2010) and Pham et al. (2019a; 2019b;2019c; 2021). On the other hand, to deal with the nonlinearity and unknown input disturbances in the dynamic model, many LMI-based methods are proposed in the literature, such as Tudon-Martinez et al. (2018); Pham et al. (2019a; 2019b; 2019c) and Vela et al. (2018). In particular, Tudon-Martinez et al. (2018) wrote the dynamic system into the LPV form with the nonlinearity considered as a scheduling variable and then used the sensors to compute the scheduling parameter, leading to an increase in the required input sensors.

To deal with the above issues, the authors have developed several estimation approaches using a nonlinear dynamic model of the semi-active damper and the two accelerometers, while the nonlinearity is bounded by a Lipschitz condition. Using two accelerometers, Pham et al. (2019a; 2019b) developed two damper force estimation methods based on the H_{∞} and unified H_{∞} observers, respectively. In these studies, the nonlinearity in the electro-rheological (ER) damper model is bounded by the Lipschitz condition. However, the results from the above nonlinear Lipschitz modeling can be conservative due to the presence of the maximum control bound in the LMI. To relax the conservatism, the dynamic system is modeled into the nonlinear parameter-varying (NLPV) representation, and an NLPV observer is then proposed to estimate the damper force by using the H_{∞} criterion to minimize the effect of unknown disturbances on the estimation error (Pham et al., 2019c; 2021).

In this work, the extended quarter-car model is represented in a quasi-LPV form, where the nonlinearity in the ER damper model is defined as the scheduling parameter. Then, a generalized H_2 /LPV observer is proposed to estimate the damping force in the presence of unknown road input and measurement noise, while the scheduling variable is obtained from estimated states. The error between the estimated scheduling variable and the real one is considered as an unknown input, which is handled using the generalized H_2 norm. The contributions of this paper are then summarized as:

- A generalized H₂/LPV observer is developed to estimate the damper force, minimizing the effect of unknown disturbances on the estimation error w.r.t the energy-to-peak gain.
- The proposed approach is simulated with a validated model of the 1/5-scaled testbed at GIPSA-lab (ANR, 2010). The simulation results demonstrate the effectiveness of the method.

The rest of the paper is organized as follows. Firstly, the quarter-car system modeling is presented in Section 2. The central part of this paper, observer formulation and design method, is detailed in Section 3. In Section 4, the observer is synthesized and analyzed in the frequency domain. To demonstrate the effectiveness of the observer, simulation results are given in Section 5. Finally, the concluding remarks are presented in Section 6.

2 Quarter-car system modeling 2.1 Physical model

Section 2.1 shows the quarter-car model augmented with the semi-active ER suspension system depicted in Fig. 1. The dynamic model includes the sprung mass (m_s) , the unsprung mass (m_{us}) , the semi-active suspension located between (m_s) and (m_{us}) , and the tire, which is modeled as a spring with stiffness k_t . From Newton's second law of motion, the system dynamics around the equilibrium are given as follows:

$$\begin{cases} m_s \ddot{z}_s = -F_s - F_d \\ m_{us} \ddot{z}_{us} = F_s + F_d - F_t \end{cases},\tag{1}$$

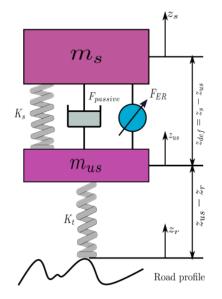


Fig. 1 1/4 car model with semi-active suspension

where $F_s = k_s(z_s - z_{us})$ is the spring force; $F_t = k_t(z_{us} - z_r)$ is the tire force; the damper force F_d is later presented in Eq. (2); z_s and z_{us} are the displacements of the sprung and unsprung masses, respectively; z_r is the road displacement input.

From Pham (2020), the nonlinear dynamical model of semi-active ER damper is given as

$$\begin{cases} F_{d} = k_{0} \left(z_{s} - z_{us} \right) + c_{0} \left(\dot{z}_{s} - \dot{z}_{us} \right) + F_{er} \\ \dot{F}_{er} = -\frac{1}{\tau} F_{er} + \frac{f_{c}}{\tau} \cdot u \cdot \tanh \left(k_{1} \left(z_{s} - z_{us} \right) + c_{1} \left(\dot{z}_{s} - \dot{z}_{us} \right) \right), \end{cases}$$
(2)

where F_d is the damper force; c_0 , c_1 , k_0 , k_1 , f_c , and τ are constant parameters. The parameters of the model in Eq. (2) are shown in Table 1. The control input u is the applied voltage that provides the electrical field to control the ER damper. In practice, it is the duty cycle of the PWM signal that controls the application (shown in Table 2).

2.2 Quasi-LPV modeling

By	selecting	the	system	states	as		
$x = \left[x_{1}, x_{2}, x_{3}, x_{4}, x_{5}\right]^{T} = \left[z_{s} - z_{us}, \dot{z}_{s}, z_{us} - z_{r}, \dot{z}_{us}, F_{er}\right]^{T} \in \mathbb{R}^{5},$							
				$y = \begin{bmatrix} \ddot{z}_s, \ddot{z}_{us} \end{bmatrix}^T \in \mathbb{R}$ ing paramete			
$\rho_1 = \tanh(k_1(z_s - z_{us}) + c_1(\dot{z}_s - \dot{z}_{us})) \in \mathbb{R}, \ \rho_2 = u, \ \text{the}$							
•	•			formulated in representatio			
$\begin{cases} \dot{x} = A \\ y = C \end{cases}$	$\mathbf{x} + \mathbf{B}(\rho_1) \cdot \rho_2 + \mathbf{B}(\rho_1) \cdot \rho_2 + \mathbf{D}_2 \boldsymbol{\omega}$	$+ \boldsymbol{D}_{1}\omega,$		(3)		

Table 1 Parameters of the quarter-car model with an ER damper

$\begin{array}{c c c c c c c c c c c } \hline m_s & Sprung mass & 2.27 & kg \\ \hline m_{us} & Unsprung mass & 0.25 & kg \\ \hline k_s & Spring stiffness & 1396 & N/m \\ \hline k_l & Tire stiffness & 12270 & N/m \\ \hline k_l & Effective stiffness coefficient due to the gas pressure & 170.4 & N/m \\ \hline c_0 & Viscous damping coefficient in the absence of control input & 68.83 & N s/m \\ \hline k_1 & Hysteresis coefficient due to displacement & 218.16 & N s/m \\ \hline c_1 & Hysteresis coefficient due to velocity & 21 & N s/m \\ \hline f_c & Dynamic yield force of ER fluid & 28.07 & N \\ \hline \tau & Time constant & 43 & ms \\ \hline \hline Table 2 Range of the control input value u \\ \hline \hline u & Duty cycle of PWM channel & [0, 1] \\ \hline \end{array}$	Parameter	Description	Value	Unit				
k_s Spring stiffness1396N/m k_i Tire stiffness12270N/m k_0 Effective stiffness coefficient due to the gas pressure170.4N/m c_0 Viscous damping coefficient in the absence of control input68.83N s/m k_1 Hysteresis coefficient due to displacement218.16N s/m c_1 Hysteresis coefficient due to velocity21N s/m f_c Dynamic yield force of ER fluid28.07N τ Time constant43msTable 2 Range of the control input value u Control inputDescriptionValue	m _s	Sprung mass	2.27	kg				
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Table 2 Range of the control input value u Control input Description Value	f_c	Dynamic yield force of ER fluid	28.07	Ν				
Control input Description Value	τ	Time constant	43	ms				
1 1	Table 2 Range of the control input value u							
<i>u</i> Duty cycle of PWM channel [0, 1]	Control inpu	ontrol input Description						
	u	Duty cycle of PWM channel						

where $\omega = \begin{pmatrix} \omega_r \\ n_y \end{pmatrix}$, in which $\omega_r = \dot{z}_r$ is the road profile derivative and n_y is the sensor noises, and the system matrices are as follows:

$$\boldsymbol{A} = \begin{bmatrix} 0 & 1 & 0 & -1 & 0 \\ \frac{-(k_s + k_0)}{m_s} & \frac{-c_0}{m_s} & 0 & \frac{c_0}{m_s} & \frac{-1}{m_s} \\ 0 & 0 & 0 & 1 & 0 \\ \frac{(k_s + k_0)}{m_{us}} & \frac{-c_0}{m_{us}} & \frac{-k_r}{m_{us}} & \frac{c_0}{m_{us}} & \frac{-1}{m_{us}} \\ 0 & 0 & 0 & 0 & \frac{-1}{\tau} \end{bmatrix},$$
$$\boldsymbol{B} \left(\rho_1 \right) = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ \frac{f_C}{\tau} \rho_1 \end{bmatrix},$$
$$\boldsymbol{C} = \begin{bmatrix} \frac{-(k_s + k_0)}{m_s} & \frac{-c_0}{m_s} & 0 & \frac{c_0}{m_s} & \frac{-1}{m_s} \\ \frac{(k_s + k_0)}{m_{us}} & \frac{-c_0}{m_{us}} & \frac{-k_r}{m_{us}} & \frac{c_0}{m_{us}} & \frac{-1}{m_{us}} \end{bmatrix},$$
$$\boldsymbol{D}_1 = \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 0 & 0 \\ -1 & 0 \\ 0 & 0 \end{bmatrix}, \quad \boldsymbol{D}_2 = \begin{bmatrix} 0 & 0.01 \\ 0 & 0.01 \\ 0 & 0.01 \end{bmatrix}.$$

3 Observer design

3.1 Quasi-LPV observer definition

In Section 3.1, a generalized H_2 /LPV observer is designed to estimate the damper force accurately.

According to the damper model in Eq. (2), the estimated force \hat{F}_d can then be obtained as

$$\hat{F}_{d} = k_0 \hat{x}_1 + c_0 \left(\hat{x}_2 - \hat{x}_4 \right) + \hat{x}_5 .$$
(4)

Therefore, the variables to be estimated are defined by the output vector $\mathbf{z} = [x_1, x_2, x_3, x_4, x_5]^T$.

Finally, the generalized H_2 //LPV observer for the system in Eq. (3) is formulated as

$$\begin{cases} \dot{\hat{x}} = A\hat{x} + B(\hat{\rho}_1) \cdot \rho_2 + L(\rho_2)(y - C\hat{x}) \\ \dot{\hat{z}} = C_z \hat{x} \end{cases},$$
(5)

where \hat{x} denotes the estimated states of the system's states x, \hat{z} denotes the estimated variables of z. The observer gain $L(\rho_2)$ is to be obtained through an optimization problem that is detailed in Section 3.2 (in the framework of quadratic stability of the estimation error).

3.2 Observer design

The state estimation error e(t) is defined as

$$e(t) = x(t) - \hat{x}(t).$$
(6)

Differentiating e(t) w.r.t time and using Eq. (3) and Eq. (6), one obtains

$$\begin{cases} \dot{e} = \dot{x} - \dot{\hat{x}} = Ax + B(\rho_1) \cdot \rho_2 + D_1 \omega \\ -A \hat{x} - L(\rho_2)(y - C \hat{x}) - B(\hat{\rho}_1) \cdot \rho_2 , \\ e_z = C_z e \end{cases}$$
(7)

equivalently,

$$\begin{cases} \dot{e} = (\boldsymbol{A} - \boldsymbol{L}(\rho_2)\boldsymbol{C})\boldsymbol{e} + (\boldsymbol{B}(\rho_1) - \boldsymbol{B}(\hat{\rho}_1)) \cdot \rho_2 \\ + (\boldsymbol{D}_1 - \boldsymbol{L}(\rho_2)\boldsymbol{D}_2)\boldsymbol{\omega} \\ \boldsymbol{e}_z = \boldsymbol{C}_z \boldsymbol{e} \end{cases}$$
(8)

In this work $\boldsymbol{B}(\rho_1) - \boldsymbol{B}(\hat{\rho}_1)$ is the inexact varying variable, which is considered as uncertainty. From the system matrix $\boldsymbol{B}(\rho)$, one obtains

$$\boldsymbol{B}(\boldsymbol{\rho}_{1}) - \boldsymbol{B}(\hat{\boldsymbol{\rho}}_{1}) = \boldsymbol{B} \cdot (\boldsymbol{\rho}_{1} - \hat{\boldsymbol{\rho}}_{1}), \qquad (9)$$

where $\boldsymbol{B} = \begin{bmatrix} 0 & 0 & 0 & f_c / \tau \end{bmatrix}^T$. Here, we assume that this uncertainty $(\rho_1 - \hat{\rho}_1)$ is bounded by a constant as

$$\left(\rho_{1}-\hat{\rho}_{1}\right)=\Delta\rho_{1}\cdot n_{\rho}, \qquad (10)$$

where $\Delta \rho_1$ is a constant matrix and n_{ρ} is white noise. Denoting $\Delta B = B \Delta \rho_1$, the dynamic estimation error system in Eq. (8) is written as

$$\begin{cases} \dot{e} = (\boldsymbol{A} - \boldsymbol{L}(\rho_2)\boldsymbol{C})\boldsymbol{e} + \Delta \boldsymbol{B}(\rho_2) \cdot \boldsymbol{n}_{\rho} + (\boldsymbol{D}_1 - \boldsymbol{L}(\rho_2)\boldsymbol{D}_2)\boldsymbol{\omega} \\ \boldsymbol{e}_z = \boldsymbol{C}_z \boldsymbol{e} \end{cases}$$
(11)

Defining a new unknown input $\omega_n = \begin{pmatrix} n_\rho \\ \omega \end{pmatrix}$, Eq. (11) is rewritten as

$$\begin{cases} \dot{e} = (\mathbf{A} - L(\rho_2)\mathbf{C})\mathbf{e} + \mathbf{D}(\rho_2)\omega_n \\ e_z = \mathbf{C}_z \mathbf{e} \end{cases},$$
(12)

where $\boldsymbol{D}(\rho_2) = \left[\Delta \boldsymbol{B}(\rho_2) (\boldsymbol{D}_1 - L(\rho_2) \boldsymbol{D}_2)\right].$

Since the system in Eq. (12) only depends on ρ_2 we can choose to define the observer gain $L(\rho_2)$ in a polytopic form as $L(\rho_2) = \sum_{i=1}^{2} \alpha_i (\rho_2) L_i$, where L_i is the observer gain at each vertex of ρ_2 and $\sum_{i=1}^{2} \alpha_i (\rho_2) = 1$. The generalized H_2 /LPV observer design objectives are:

- the system in Eq. (12) is stable for $\omega_n = 0$;
- $\frac{\|e_z\|_{\infty}}{\|\omega_n\|_2}$ is minimized for $\omega_n \neq 0$.

The following theorem solves the above problem following an LMI framework (Scherer and Weiland, 2015).

Theorem 1: Consider the system model in Eq. (3) and the observer in Eq. (5). If there exists a symmetric positive definite matrix **P** and matrices Y_i with i = 1, 2 minimizing γ such that

$$\begin{pmatrix} \boldsymbol{P}\boldsymbol{A} + \boldsymbol{A}^{T}\boldsymbol{P} - \boldsymbol{Y}_{i}\boldsymbol{C} - (\boldsymbol{Y}_{i}\boldsymbol{C})^{T} & \boldsymbol{P}\Delta\boldsymbol{B}(\boldsymbol{\rho}_{2i}) & \boldsymbol{P}\boldsymbol{D}_{1} - \boldsymbol{Y}_{i}\boldsymbol{D}_{2} \\ \Delta\boldsymbol{B}(\boldsymbol{\rho}_{2i})^{T}\boldsymbol{P} & -\boldsymbol{I} & \boldsymbol{0} \\ \boldsymbol{D}_{1}^{T}\boldsymbol{P} + (\boldsymbol{Y}\boldsymbol{D}_{2})^{T} & \boldsymbol{0} & -\boldsymbol{I} \end{pmatrix} < \boldsymbol{0}, \\ \begin{pmatrix} \boldsymbol{P} & \boldsymbol{C}_{z}^{T} \\ \boldsymbol{C}_{z} & \boldsymbol{\gamma}\boldsymbol{I} \end{pmatrix} > \boldsymbol{0}, \end{cases}$$
(13)

then the observer vertex matrices L_i determined from $L_i = P^{-1}Y_i$ ensure that the objectives are attained. I is the identity matrix.

Proof. The proof is not detailed here since it consists in a simple application of the generalized H_2 condition to each vertex of the dynamic estimation error system in Eq. (12).

4 Observer synthesis and frequency domain analysis 4.1 Observer synthesis

As previously mentioned, the scheduling variable $\rho_1 = \tanh(k_1x_1 + c_1(x_2 - x_4))$ is limited in the range of [-1, 1] and $\rho_2 = u$ varies in the range of [0, 1]. Therefore, the following values are obtained:

- the lower bound of ρ_2 is $\rho_2 = 0$;
- the upper bound of ρ_2 is $\overline{\rho}_2 = 1$.

The Yalmip toolbox (Lofberg, 2004) and the sdpt3 solver (Toh et al., 1999) are used to solve the optimization in Theorem 1. Solving Theorem 1 with the two above vertices leads to the minimum L_2 -induced gain $\gamma = 0.478$, and to the observer vertex matrices

$$\boldsymbol{L}_{1} = \begin{bmatrix} -0.1011 & 0 \\ -0.0093 & -0.0002 \\ 32.7307 & -327.9806 \\ -0.1009 & 1 \\ -128.9226 & -0.3546 \end{bmatrix},$$

	-0.0998	0.0002 -	
	0.0018	0.0014	
$L_{2} =$	32.7413	-327.9807	
	-0.0998	0.2794	
	-128.0171		

4.2 Frequency-domain analysis

In Section 4.2, the analysis of the effects of unknown inputs (road profile derivative ω_r and measurement noises *n*) on the estimation error *e* in the frequency domain is performed using the Bode diagrams.

The transfer functions from ω_r and *n* to the estimation error e_z are shown on the left and right sides of Fig. 2. As shown, the Bode diagrams of the polytopic at the frozen values of two vertices ($\rho_2 = 0$ (red line) and $\bar{\rho}_2 = 1$ (green dash-dot dot)) emphasizes the satisfactory attenuation level of the unknown disturbances on the estimation error.

5 Simulation results

In Section 5, simulations with different scenarios are performed to validate the designed observer in the time domain.

5.1 Simulation scenario 1

This simulation scenario is as follows:

- The road profile input is a chirp signal with an amplitude of 7×10^{-3} m and various frequencies from 0 Hz to 10 Hz (shown in Fig. 3).
- The scheduling parameter ρ_2 ($\rho_2 = u$) is obtained from a Skyhook controller (see Fig. 4).

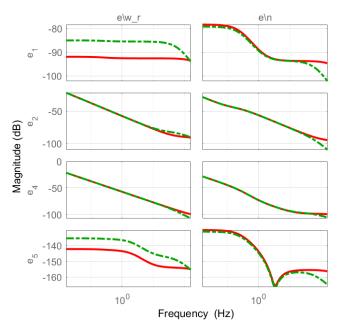
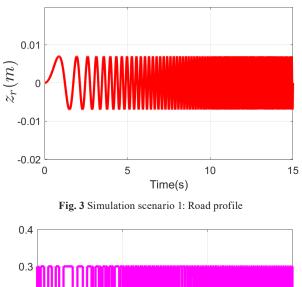


Fig. 2 Bode diagram w.r.t road profile derivative (left) and measurement noise (right)



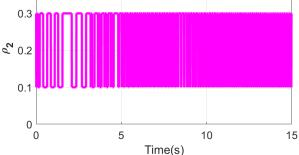


Fig. 4 Simulation scenario 1: Scheduling parameter ρ_2

The estimated scheduling variable $\hat{\rho}_1$ is shown in Fig. 5. Fig. 6 shows the simulated force in the solid red line and the estimated force in the blue dotted line, respectively. The estimation error is shown in Fig. 7. We can see that the estimation error converges to 0 after 1 second. Therefore, the proposed method is stable with the various frequencies of road profile disturbance.

5.2 Simulation scenario 2

In the second simulation scenario, the proposed observer is validated with more realistic road profile.

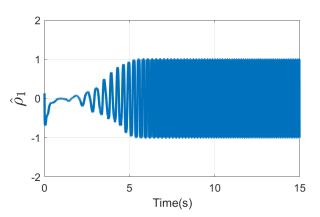


Fig. 5 Simulation scenario 1: Estimated scheduling parameter $\hat{\rho}_1$

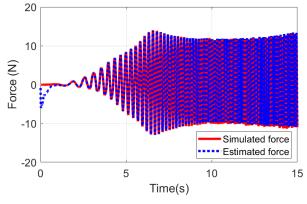


Fig. 6 Simulation scenario 1: Real and estimated damper force

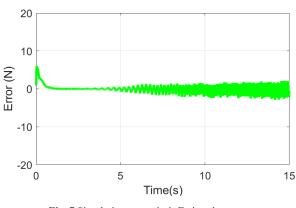


Fig. 7 Simulation scenario 1: Estimation error

The second simulation is designed as follows:

- A realistic road profile (ISO type C) is used (shown in Fig. 8).
- The scheduling parameter ρ₂ (ρ₂ = u) comes from a Skyhook controller (shown in Fig. 9).

The estimated scheduling parameter $\hat{\rho}_1$ is shown in Fig. 10. The simulation results of this scenario are shown in Figs. 11 and 12. The results show the stability of the proposed schemes $e \rightarrow 0$ in a realistic case. It is worth noting that the observer must work when the sensors are affected

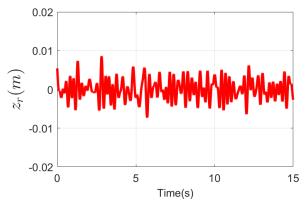


Fig. 8 Simulation scenario 2: ISO road profile (Type C)

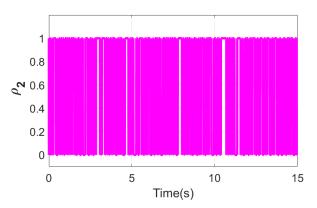


Fig. 9 Simulation scenario 2: Scheduling parameter ρ_2

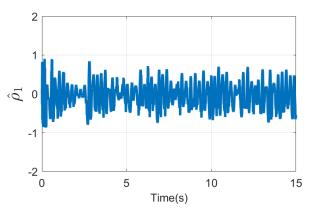


Fig. 10 Simulation scenario 2: Estimated scheduling parameter $\hat{\rho}_1$

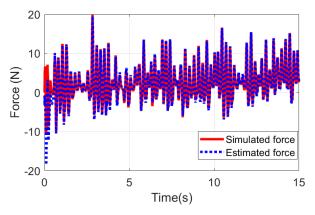


Fig. 11 Simulation scenario 2: Real and estimated damper force

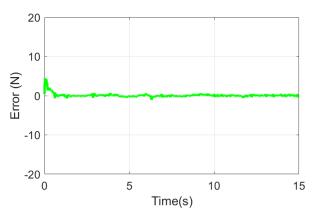


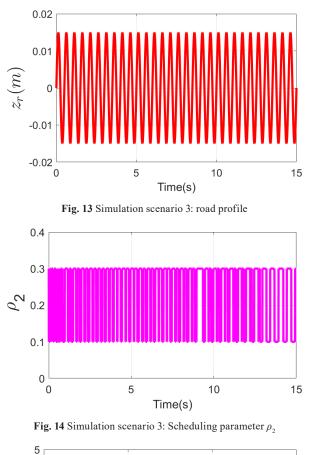
Fig. 12 Simulation scenario 2: Estimation error

by noises in the real application. Therefore, to assess the effectiveness of the observer affected by noises, a more realistic scenario is presented in Section 5.3.

5.3 Simulation scenario 3

This scenario is designed by adding the noises into the measurement outputs, detailed as follows:

- A sinusoidal road profile is considered in this scenario (shown in Fig. 13).
- The scheduling parameter ρ₂ is obtained from a Skyhook controller (Fig. 14).
- The sensors are affected by noises (shown in Figs. 15 and 16).



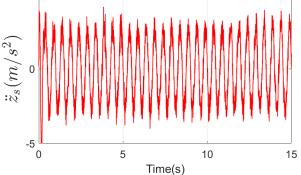
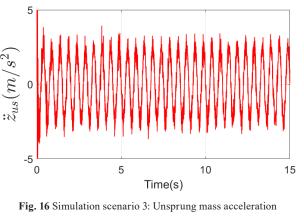


Fig. 15 Simulation scenario 3: Sprung mass acceleration affected by noise

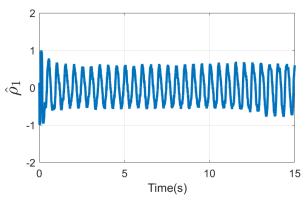


affected by noise

The estimated scheduling parameter $\hat{\rho}_1$ is presented in Fig. 17. The damper estimation results of this scenario are shown in Fig. 18 while the estimation error is seen in Fig. 19. The results present the robustness of the proposed approaches in the presence of the sensor noises.

6 Conclusion

This paper presents a generalized H_2 /LPV observer to estimate the damper force. For this purpose, the quarter-car system is formulated in a quasi-LPV form. A generalized H_2 /LPV observer using accelerometer measurements as input is developed, providing an accurate estimation of





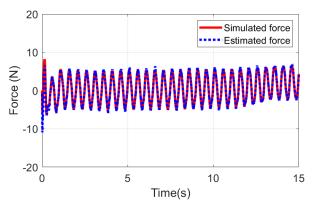


Fig. 18 Simulation scenario 3: damper force

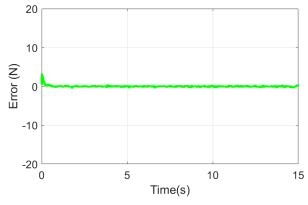


Fig. 19 Simulation scenario 3: Estimation error

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the damper force in real-time. The force estimation error is minimized, accounting for the effect of unknown inputs (road profile derivative measurement noise). The simulation results show the performance and the accuracy of the proposed approach.

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